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# MODIFIED BOOTH MULTIPLIER ARCHITECTURE USING NEW (1, 1, 1) ADDER

Anu Mehra, Priyank Kularia, Aditya Sharma, Garima Batra, Achintya Rawat and Nidhi Gaur  
Amity School of Engineering and Technology Amity University, Noida, India  
E-Mail: [amehra@amity.edu](mailto:amehra@amity.edu)

## ABSTRACT

In this paper an alternate implementation of the modified Booth algorithm is presented where groups of the partial product terms are summed using parallel prefix adders proposed by Harris *et al.* Comparative analysis of these adders in terms of power, delay and LUTs is performed. A modified 16 bit multiplication process using Radix 4 Booth Algorithm is proposed and results with respect to Kogge Stone and New (1, 1, 1) adder are computed. Simulation results are carried out on Xilinx Vivado Version 14.2 on Artix 7 Board.

**Keywords:** Kogge Stone Adder, modified booth architecture, adder.

## INTRODUCTION

### Multiplier

Multipliers are essential hardware for designing digital circuits. These systems include digital signal processors, microprocessors etc. High speed and low power consumption are the most important attributes required for high performance of multipliers [1]. Booth multiplication algorithm uses small number of additions and shift operations to perform multiplication [2]. Therefore, Booth multipliers have been a part of High speed multipliers and several modifications and applications have been proposed in literature [3].

Radix 2 involves scanning of 2 bits at a time for multiplication. The pioneer version of Booth's multiplier (Radix-2) has two limitations:

1. Fluctuation in the number of add or subtract operations is observed.
2. The algorithm becomes inconvenient in the case of isolated 1's.

In Radix-4 algorithm [4], number of steps is halved for the calculation of partial products, since scanning of 3 bits at a time is performed as compared to Radix 2 where 2 bits are scanned at a time.

### Parallel Prefix Adder

Parallel prefix adders [5] [6] take inputs  $\{A_{N-1}, \dots, A_0\}$ ,  $\{B_{N-1}, \dots, B_0\}$  and  $C_{IN}$ , and produce a sum output  $\{S_N, \dots, S_1\}$  using intermediate Generate (G) and propagate (P) prefix signals. The addition logic consists of the following 3 stages: [7]

### Pre-processing

This step involves computation of generate and propagate signals corresponding to each pair of bits in A and B. These signals are given by the logic equations below:

$$P_i = A_i \text{ XOR } B_i \\ G_i = A_i \text{ AND } B_i = A_i B_i$$

### Carry Look-ahead network

This step involves computation of carries corresponding to each bit. It uses group propagate and generate as intermediate signals which are given by the logic equations below:

$$P_0 = A_0 \text{ XOR } B_0 \\ G_0 = A_0 \text{ AND } B_0 = C_0 \\ \text{Sunsequent carries } C_1, C_2, C_3 \text{ etc. are} \\ C_1 = G_0 + P_0 C_0 \\ C_2 = G_1 + P_1 C_1 \\ C_3 = G_2 + P_2 C_2 \text{ etc. In general,} \\ C_i = G_{i-1} + P_{i-1} C_{i-1}$$

Substituting  $C_1, C_2$  etc. in subsequent equations gives

$$C_2 = G_1 + P_1 G_0 + C_0 P_0 P_1 \\ C_3 = G_2 + G_1 P_2 + G_0 P_1 P_2 + C_0 P_0 P_1 P_2 \\ C_4 = G_3 + G_2 P_3 + G_1 P_2 P_3 + G_0 P_1 P_2 P_3 + C_0 P_0 P_1 P_2 P_3 \text{ etc.}$$

### Post processing

This is the final step and is common to all adders of this family. It involves computation of sum bits. Sum bits are computed by the logic given below:  $S_i = P_i \text{ XOR } C_{i-1}$

$$P_{ij} = P_{i:k+1} \text{ AND } P_{k:j} \\ \text{e.g. } P_{3:0} = P_3 P_2 P_1 P_0 \\ P_{3:2} = P_3 P_2 \\ G_{ij} = G_{i:k+1} + (P_{i:k+1} \text{ AND } G_{k:j}) \\ \text{e.g. } G_{3:0} = G_{3:2} \text{ OR } (P_{3:2} \text{ AND } G_{1:0}) \\ = (G_3 + P_3 G_2) + (P_3 P_2 (G_1 + P_1 G_0))$$

Parallel Prefix adders include two basic cells:



**Black cell**

As shown in Figure-1(a) this cell is used in the logic design of adders and it produces intermediate generate and propagate terms of present and previous states.

$$G_{i,j} = G_{i,k} \text{ OR } (P_{i,k} \text{ and } G_{k-1,j})$$

$$P_{i,j} = P_{i,k} \text{ AND } P_{k-1,j}$$

**Grey cell**

As shown in Figure-1(b) this cell is used in the logic design of adders and it produces intermediate generate of present and previous states.

$$G_{i,j} = G_{i,k} \text{ OR } (P_{i,k} \text{ and } G_{k-1,j})$$

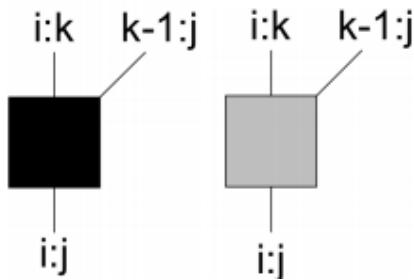


Figure-1. (a) Black cell; (b) Grey cell.

**Contributions**

We have proposed a modified Booth multiplier which reduces the delay to a greater extent for 16 bit multiplication. This makes the circuit faster and appealing to high speed digital electronic designs. The principle and working of modified approach is described in detail in Part II.

**Organization of Paper**

Paper comprised of five sections including introduction. Section II describes our technique Modified Booth Multiplier Architecture. In section III block diagram of our design is presented with explanation of each block. In section IV simulation results are analysed and compared with existing results. Conclusion of paper is discussed in section IV.

**MODIFIED BOOTH ARCHITECTURE (MBA)**

This includes dividing the multiplier into three digits correspondingly on the basis of the Booth encoder Table (Table-1). In Radix-4 booth encoder, first step is to append extra '0' to the Least Significant Bit (LSB) of the multiplier. When we add an extra '0' it becomes a 9 bit number. Now starting from the Least Significant Bit, three bits are consequently grouped e.g. first group is '0' and A<sub>7</sub> and A<sub>6</sub> and second group is A<sub>6</sub> and A<sub>5</sub> and A<sub>4</sub> where and represents concatenation. After checking the sequence from Table-1, the relevant encoding is chosen.

Table-1. Radix-4 booth algorithm encoding.

BN+1	B	BN-1	Operation	1M	2M	3M
0	0	0	0	1	1	0
0	0	1	1 * M	0	1	0
0	1	0	1 * M	0	1	0
0	1	1	2 * M	1	0	0
1	0	0	-2 * M	1	0	1
1	0	1	-1 * M	0	1	1
1	1	0	-1 * M	0	1	1
1	1	1	0	1	1	0

**BLOCK DIAGRAM OF MBA**

MBA consists of six blocks as shown in Figure-2: Sign Extension Generator, Control Unit, and Partial Product Generator.

**SIGN EXTENSION GENERATOR**

It comprises of sign extension chip and sign control chip which is clock based. Sign extension corrector improves the capability of the booth multiplier by not only multiplying the unsigned numbers but also the signed numbers.

Sign Extension Control Unit (CHIP6) is used to control the result from CHIP5 and stores them into particular and correct order and then fed the result to partial product rearrange unit.

Table-2. Sign extension table when a<sub>7</sub>=0/1.

A7	Bn+1	Bn	Bn-1	E(A7=0)	E(A7=1)
0	1	0	0	0	0
0	1	0	0	1	1
0	1	0	1	0	1
0	1	0	1	1	1
0	1	1	0	0	0
0	1	1	0	1	0
0	1	1	1	0	0
0	1	1	1	1	0

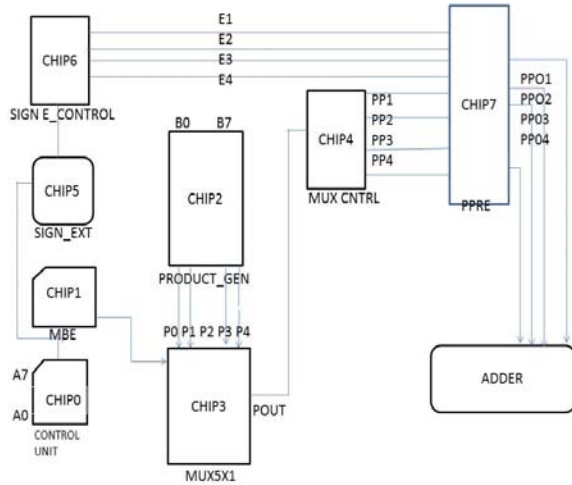


Figure-2. Block diagram of proposed multiplier.

**CONTROL UNIT**

The control unit (CHIP0) automatically fetches the combination form the multiplier and delivers it to MBE. MBE then decide on the base of result of control unit which operation is to be carried out.

**PARTIAL PRODUCT GENERATOR**

Partial product generator block consists of CHIP2, CHIP3 and CHIP4 of the MBA. Product generator (CHIP2) is structured to produce the multiple products of multiplicand which are  $0*K$ ,  $1*K$ ,  $2*K$ ,  $-1*K$  and  $-2*K$ . The product generator helps to generate the multiplicands by-product and to make decision of choosing partial product faster.

**ADDER ARCHITECTURES**

**Kogge Stone Adder**

Kogge stone adder (Figure-3) is a type of parallel prefix adder. It is a highly used adder in industries as it is considered the fastest adder which provides high performance. It gives lower fan-out. This helps in improving performance. The limitation of this adder is that it ingests large silicon area [8].

**New(1,1,1) Adder**

The parallel prefix designs can be sorted with the help of three dimensional distribution in l,f and t, where l represents the number of logic levels( logic levels = L+1), f means fanout ( $2^f + 1$ ) and t corresponds to wiring tracks ( $2^t$ ). This adder is a new member to the family of parallel prefix adders. Here l, f and t are greater than 0 [9].

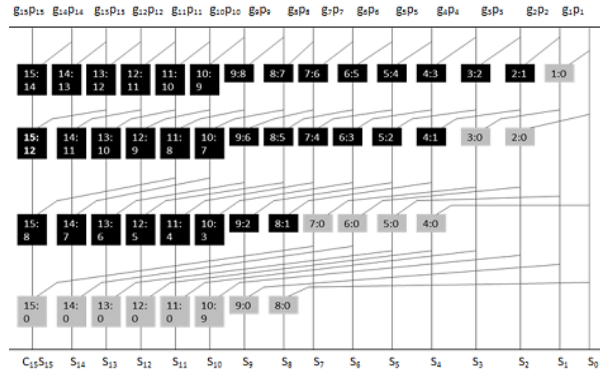


Figure-3. Kogge Stone Adder architecture.

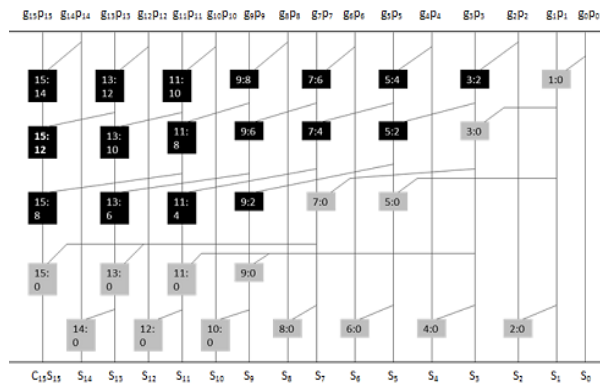


Figure-4. New (1, 1, 1) adder architecture.

**SIMULATION RESULTS**

Radix 4 Modified Booth Architecture (MBA) was implemented first using Kogge Stone which is more prevalent and then, using New (1, 1, 1) adder. The simulations were carried out in Xilinx VIVADO version 14.2 on Artix 7 Board. Comparative analysis between them is given in Table-2. Table-3 depicts the comparison of adder architectures which are implemented in the heart of Multiplier architectures. Simulation waveforms for Kogge Stone Multiplier architecture is given in Figure-5. Figure-6 gives simulation waveform of MBA multiplier.

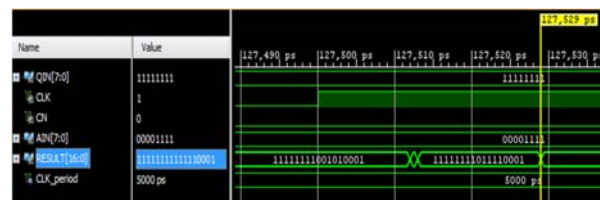


Figure-5. MBA using new (1, 1, 1) adder.

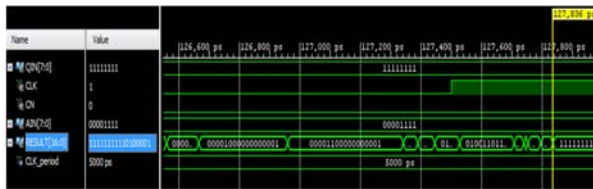


**Table-3.** Comparison of modified booth architecture using Kogge Stone and new (1, 1, 1) adder.

Multi with adder	LUTs	registers	Delay (ns)	Power (mW)		
				Total	Total	Total
m-Kogge stone	101	100	127.83	164	164	164
m-new(1,1,1)	92	100	127.52	165	165	165

**Table-4.** Comparison of Kogge Stone and new (1, 1, 1) adder architectures.

Adder	LUTs	registers	Delay (ns)	Power (mW)		
				Total	Total	Total
m-Kogge stone	46	50	5.916	163	43	120
m-new(1,1,1)	32	50	6.042	163	43	120



**Figure-6.** Multiplier using Kogge Stone Adder.

## CONCLUSIONS

From simulation results as shown in Table-3, number LUTs used in implementation of MBA with New (1, 1, 1) Adder is about 9% lower as compared to Booth using Kogge Stone. The power consumption and delay are comparable in both cases with MBA being slightly faster. Implementation with parallel processing in larger CPU etc. will lead to lesser area without degrading the results of Booth using Kogge Stone.

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